



SGM6600

90% Efficient Synchronous Step-Up Converter with 1.2A Switch

GENERAL DESCRIPTION

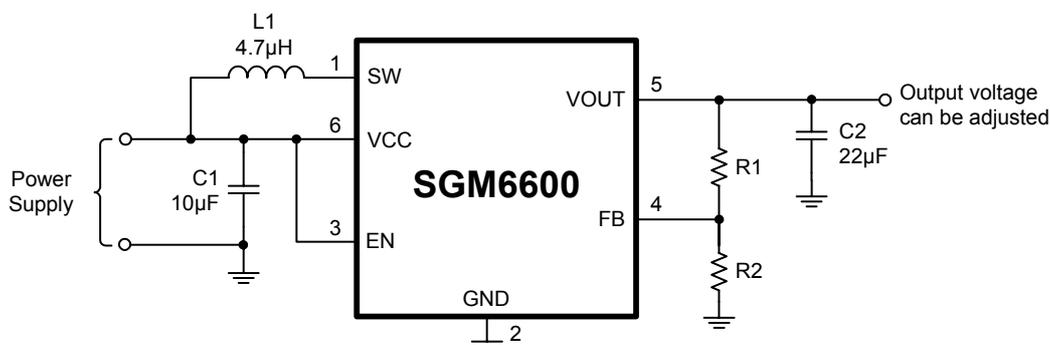
The SGM6600 is a 1.2MHz, constant frequency, current mode, synchronous, step-up switching regulator. Its Output currents can go as high as 75mA while using a single-cell alkaline, and discharge it down to 0.9V. It can also be used for generating 5V at 300mA from a 3.3V rail or a Li-ion battery.

High switching frequency minimizes the sizes of inductor and capacitor. Integrated power MOSFETs and internal compensation make the SGM6600 simple to use and fit the total solution in a compact space.

For light load current, the SGM6600 enters into the power saving mode to maintain high efficiency. Antiringing control circuitry reduces EMI concerns by damping the inductor in discontinuous mode. The SGM6600 provides true output disconnect and this allows VOUT to go to zero volts during shutdown without drawing any current from the input source.

The output voltage of SGM6600-ADJ can be programmed by an external resistor divider, and that of SGM6600-3.3/SGM6600-5.0 are fixed internally on the chip. The device is available in SOT-23-6L package. It operated over an ambient temperature range of -40°C to +85°C.

TYPICAL APPLICATION



FEATURES

- **90% Efficient Synchronous Boost Converter**
- **75mA Output Current at 3.3V from 0.9V Input**
- **150mA Output Current at 3.3V from 1.8V Input**
- **Device Quiescent Current: 200µA (TYP)**
- **Lower than 1µA in Shutdown Status**
- **Input Voltage Range: 0.9V to 5.5V**
- **3.3V and 5.0V Fixed Output Voltage**
- **Adjustable Output Voltage Up to 5.5V**
- **Power-Save Mode Version Available for Improved Efficiency at Low Output Power**
- **Load Disconnect During Shutdown**
- **Over Temperature Protection**
- **Available in Green SOT-23-6L Package**
- **-40°C to +85°C Operating Temperature Range**

APPLICATIONS

Single-Cell Li Battery Powered Products
 Portable Audio Players
 Cellular Phones
 Personal Medical Products

SGM6600

90% Efficient Synchronous Step-Up Converter with 1.2A Switch

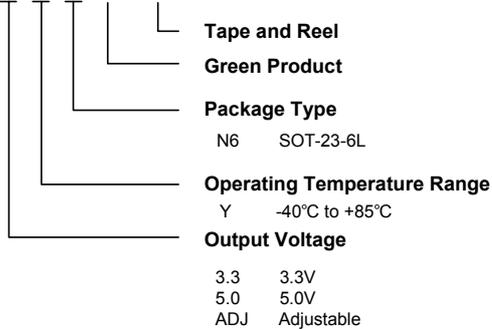
PACKAGE/ORDERING INFORMATION

MODEL	V _{OUT} (V)	PIN-PACKAGE	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKAGE OPTION
SGM6600	Adjustable	SOT-23-6L	-40°C to +85°C	SGM6600-ADJYN6G/TR	S44XX	Tape and Reel, 3000
	3.3V	SOT-23-6L	-40°C to +85°C	SGM6600-3.3YN6G/TR	S45XX	Tape and Reel, 3000
	5.0V	SOT-23-6L	-40°C to +85°C	SGM6600-5.0YN6G/TR	S46XX	Tape and Reel, 3000

NOTE: Order number and package marking are defined as the follow:

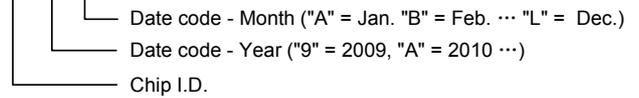
ORDER NUMBER

SGM6600-X X X G / TR



MARKING INFORMATION

SYY X X



For example: S449A (2009, January)

ABSOLUTE MAXIMUM RATINGS

Input Supply Voltage on SW, VOUT, VCC, FB	-0.3V to 6V
EN Voltage	-0.3V to (V _{OUT} + 0.3V)
Operating Temperature Range	-40°C to +85°C
Junction Temperature	150°C
Package Thermal Resistance	
SOT-23-6L, θ_{JA}	250°C/W
Storage Temperature	-65°C to +150°C
Lead Temperature (soldering, 10s)	260°C
ESD Susceptibility	
HBM	4000V
MM	200V

NOTE:

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

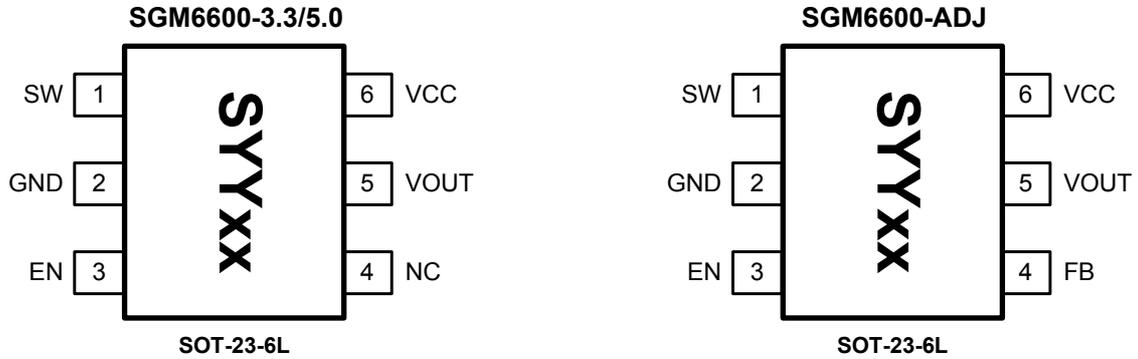
CAUTION

This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

SGMICRO reserves the right to make any change in circuit design, specification or other related things if necessary without notice at any time. Please contact SGMICRO sales office to get the last datasheet



PIN CONFIGURATIONS (Top View)



NOTE: The location of pin 1 on the SGM6600 is determined by orienting the package marking as shown.

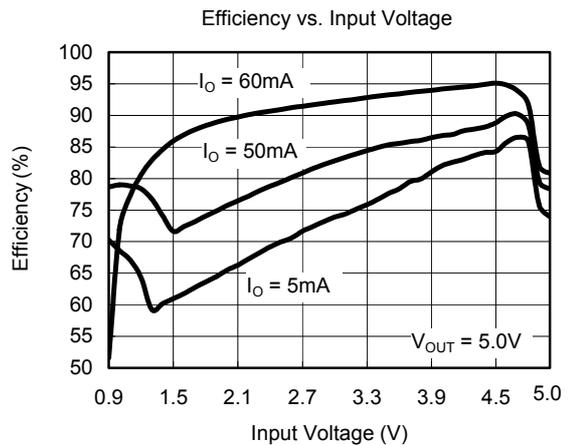
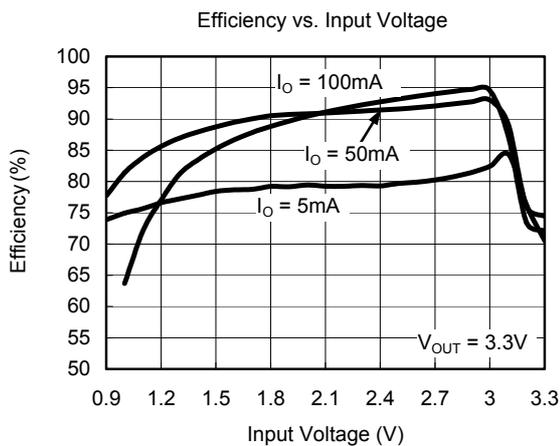
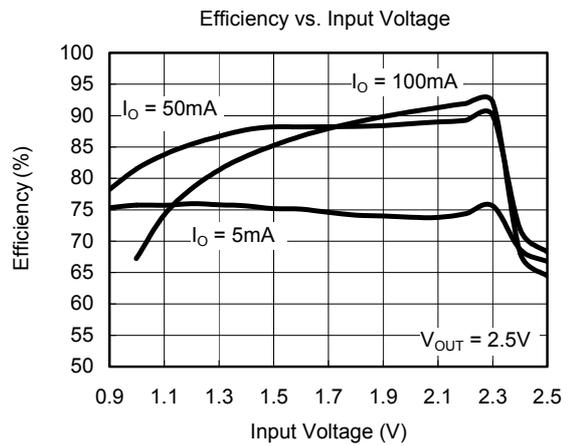
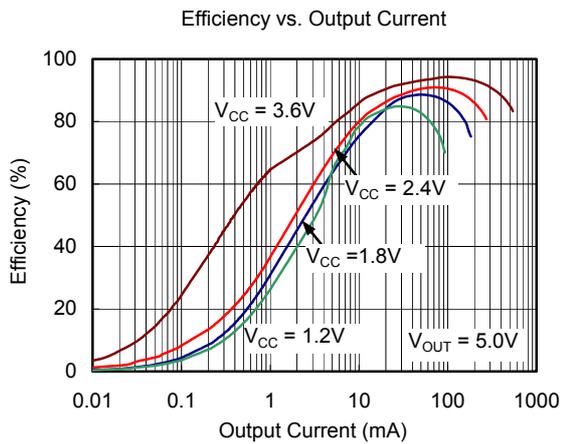
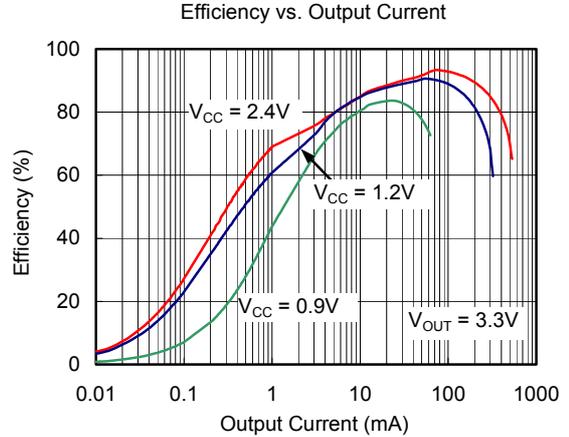
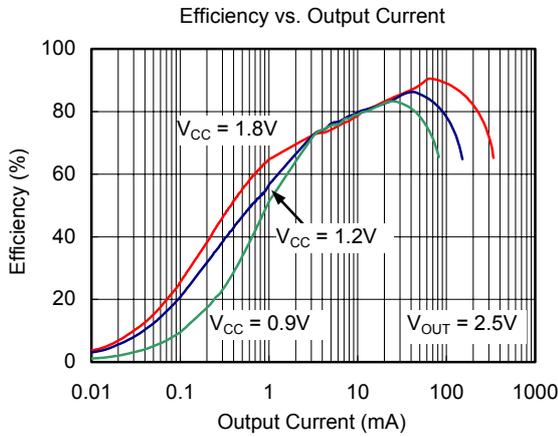
PIN DESCRIPTION

PIN	NAME	FUNCTION
1	SW	Boost and rectifying switch input.
2	GND	Ground
3	EN	Enable Input. (1/VCC enabled, 0/GND disabled)
4	NC	No Connect. It should be floating. (SGM6600-3.3/SGM6600-5.0)
	FB	Output Voltage Feedback Pin. Voltage feedback for programming the output voltage. (SGM6600-ADJ)
5	VOUT	Boost Converter Output.
6	VCC	Boost Converter Supply Voltage.

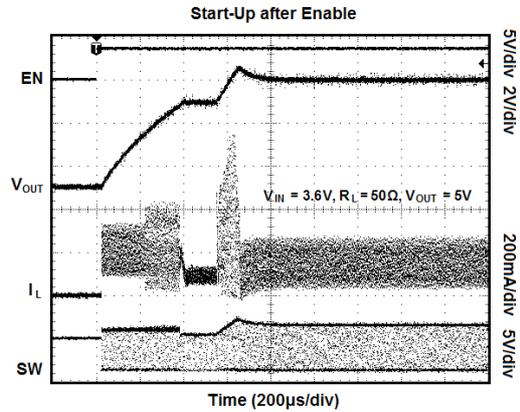
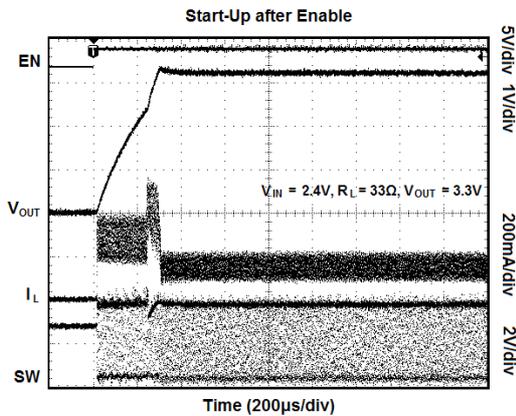
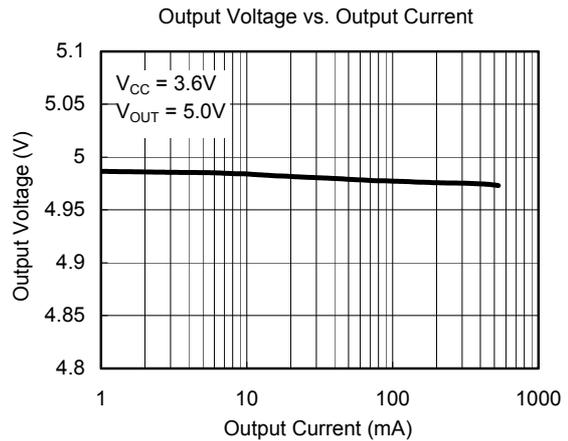
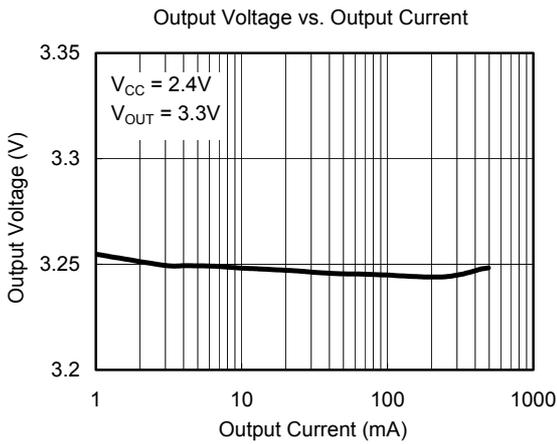
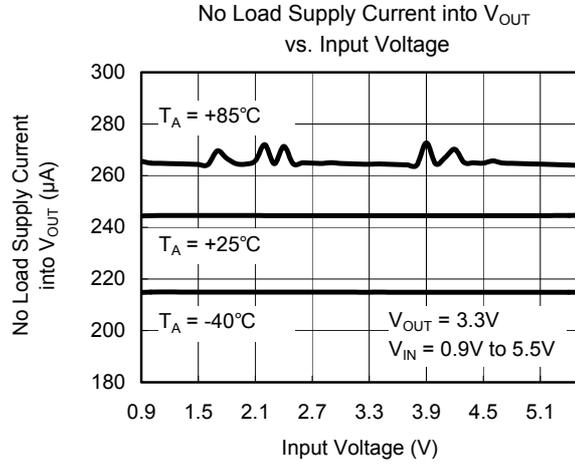
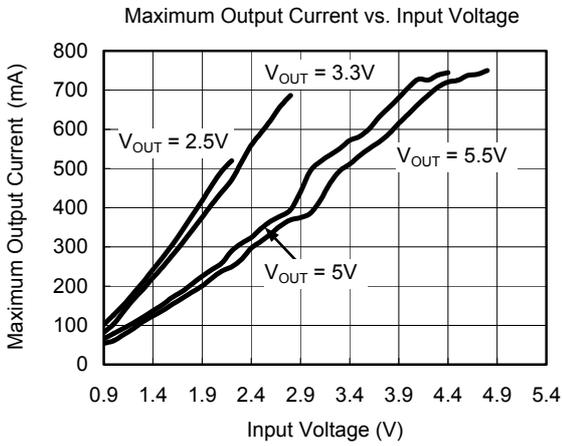
ELECTRICAL CHARACTERISTICS(Typical values are at $T_A = +25^\circ\text{C}$, Full = -40°C to $+85^\circ\text{C}$, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DC/DC STAGE						
Output Voltage Range	V_{OUT}		2.5		5.5	V
Minimum Input Voltage Range for Start-Up	V_{IN}	$R_L = 270\Omega$		1.1		V
Input Voltage Range after Start-Up	V_{IN}	$T_A = +25^\circ\text{C}$	0.9		5.5	
Feedback Voltage	V_{FB}			500		mV
Oscillator Frequency	f		900	1200	1400	kHz
Switch Current Limit	I_{SW}		950	1200	1900	mA
Start-up Current Limit				240		mA
Boost Switch-On Resistance		$V_{OUT} = 3.3\text{V}$		480		$m\Omega$
Rectifying Switch-On Resistance		$V_{OUT} = 3.3\text{V}$		600		$m\Omega$
Total Accuracy (including line and load regulation)					5	%
Line Regulation					1	%
Load Regulation					1	%
Quiescent Current	V_{CC}	$I_O = 0\text{mA}$, $V_{EN} = V_{CC} = 1.2\text{V}$,		0.1	1	μA
	V_{OUT}	$V_{OUT} = 3.3\text{V}$, $T_A = +25^\circ\text{C}$		200	235	μA
	V_{OUT}	$V_{OUT} = 5.0\text{V}$, $T_A = +25^\circ\text{C}$		230		μA
Shutdown Current		$V_{EN} = 0\text{V}$, $V_{CC} = 1.2\text{V}$, $T_A = +25^\circ\text{C}$		0.1	1	μA
CONTROL STAGE						
EN Input Low Voltage	V_{IL}	$0.9\text{V} \leq V_{CC} \leq 1.8\text{V}$			$0.2 \times V_{CC}$	V
		$1.8\text{V} < V_{CC} \leq 3.3\text{V}$			0.6	
		$3.3\text{V} < V_{CC} \leq 4.2\text{V}$			0.6	
		$4.2\text{V} < V_{CC} \leq 5.0\text{V}$			0.6	
EN Input High Voltage	V_{IH}	$0.9\text{V} \leq V_{CC} \leq 1.8\text{V}$	1.5			V
		$1.8\text{V} < V_{CC} \leq 3.3\text{V}$	2.0			
		$3.3\text{V} < V_{CC} \leq 4.2\text{V}$	2.4			
		$4.2\text{V} < V_{CC} \leq 5.0\text{V}$	2.6			
EN Input Current		Clamped on GND or VCC		0.01	0.05	μA
Overtemperature Protection				150		$^\circ\text{C}$
Overtemperature Hysteresis				20		$^\circ\text{C}$

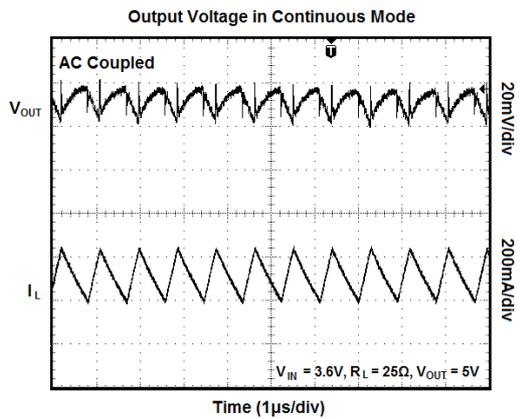
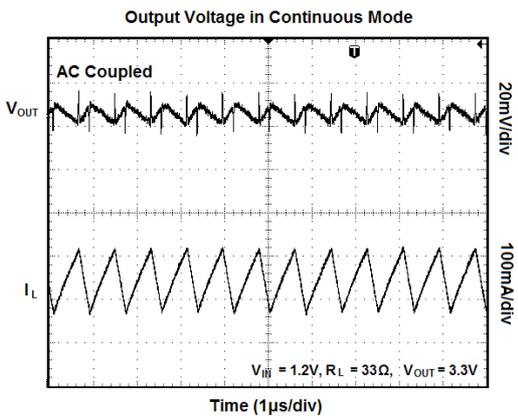
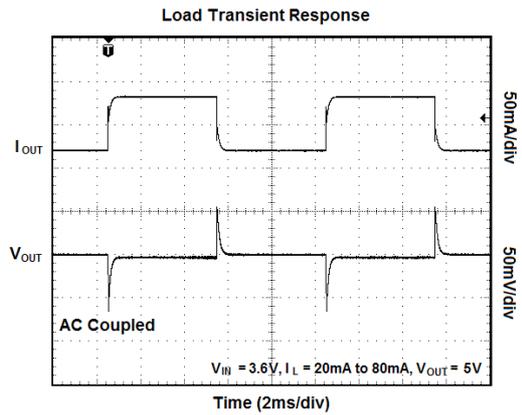
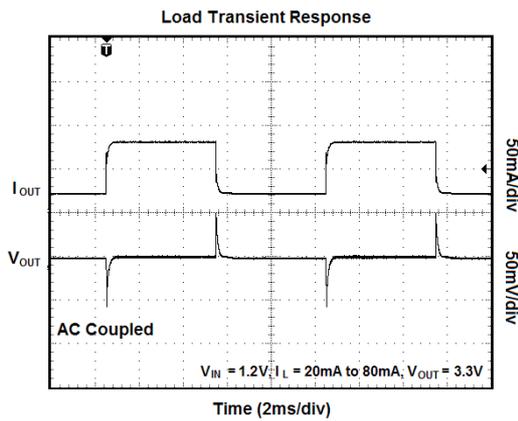
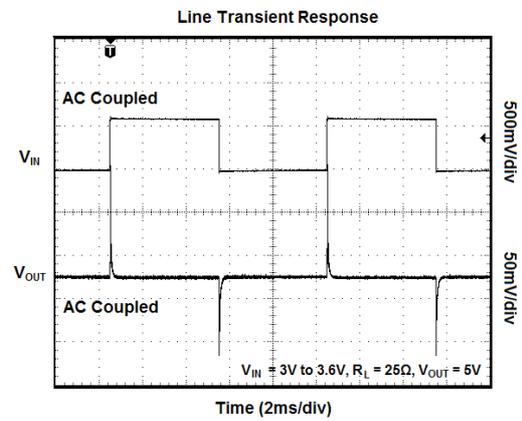
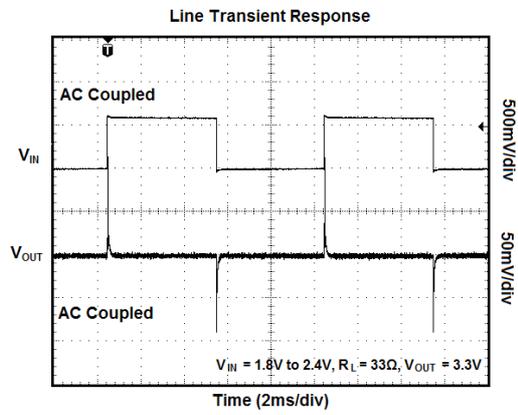
TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL APPLICATION CIRCUITS

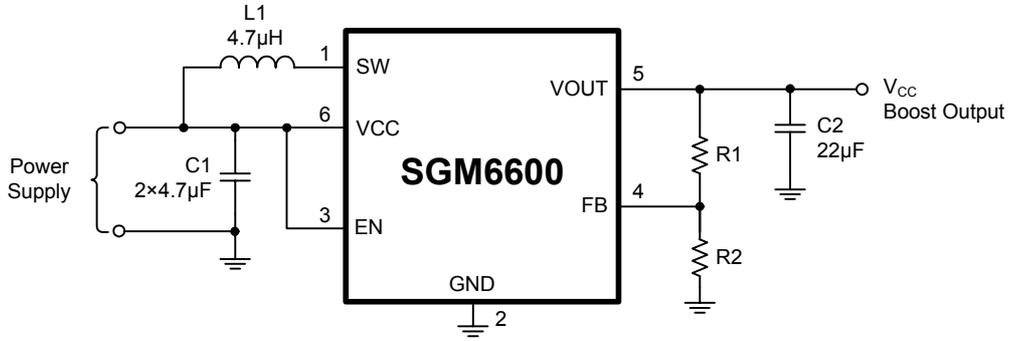


Figure 1. Power Supply Solution for Maximum Output Power Operating from a Single or Dual Alkaline Cell

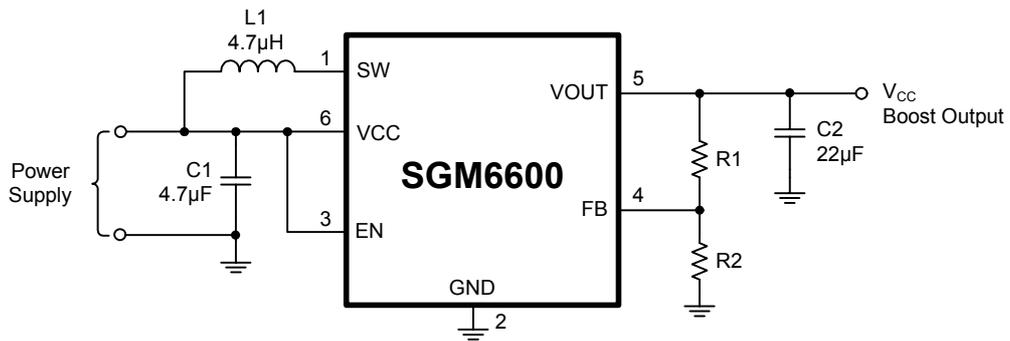


Figure 2. Power Supply Solution Having Small Total Solution Size

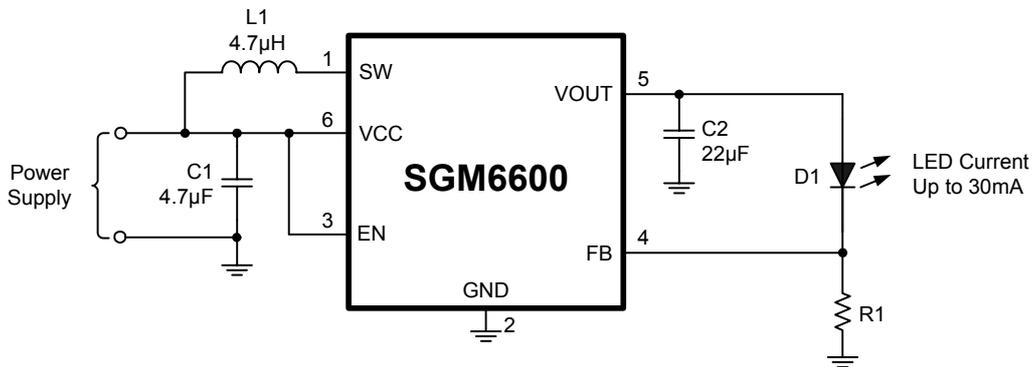


Figure 3. Power Supply Solution for Powering White LEDs in Lighting Applications

TYPICAL APPLICATION CIRCUITS

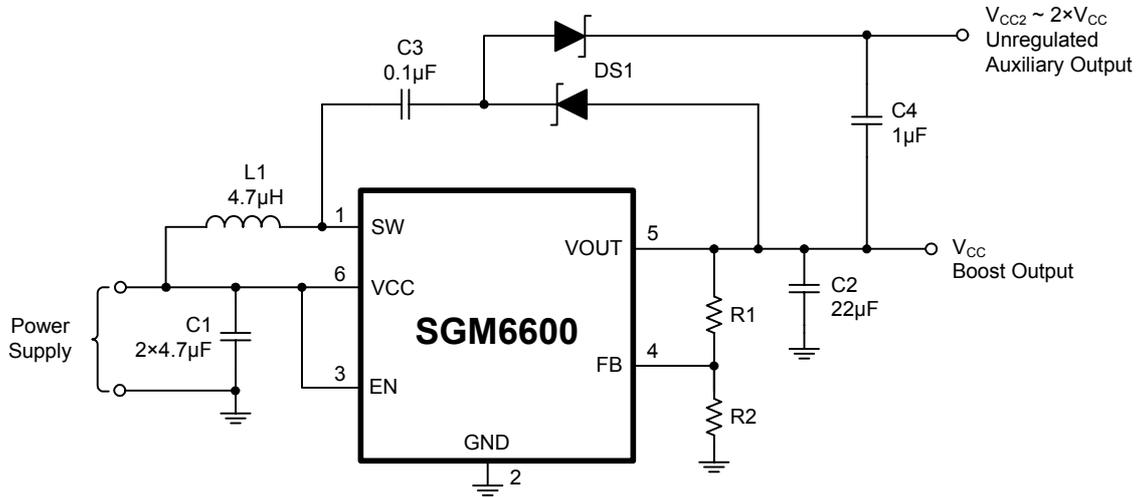


Figure 4. Power Supply Solution with Auxiliary Positive Output Voltage

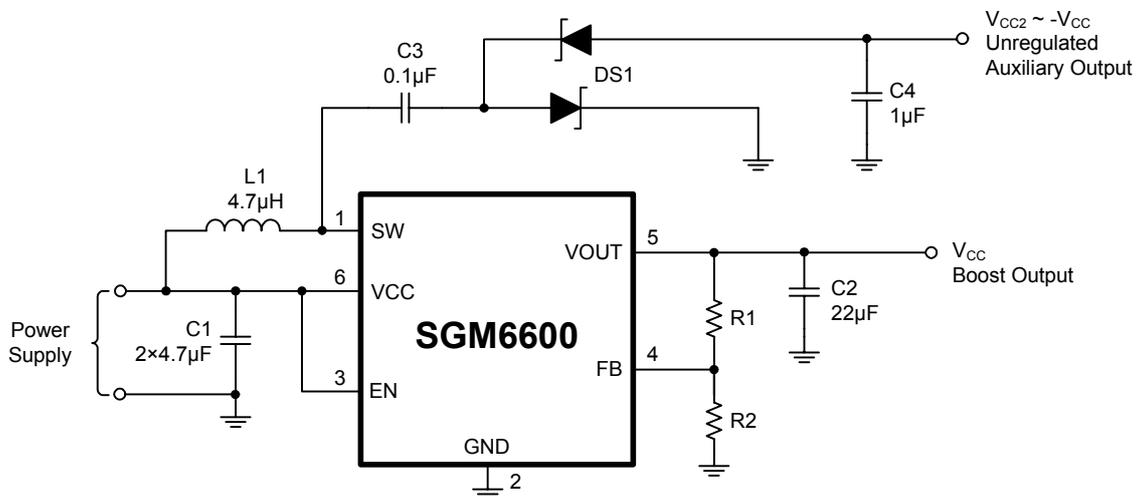


Figure 5. Power Supply Solution with Auxiliary Negative Output Voltage

TYPICAL APPLICATION CIRCUITS

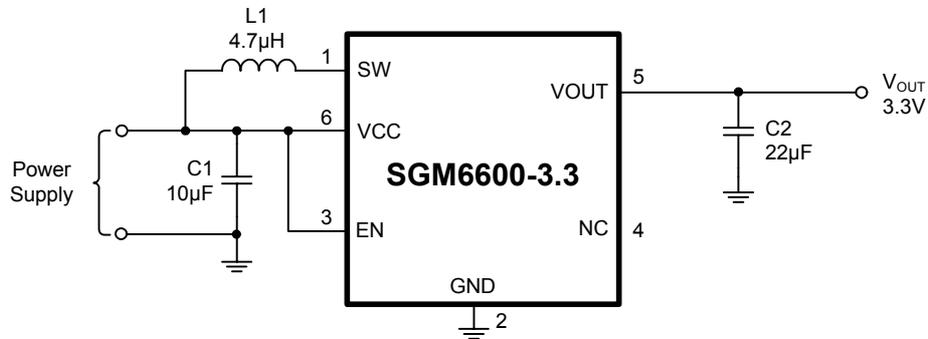


Figure 6a. Basic Application Circuit for the Fixed Output Versions

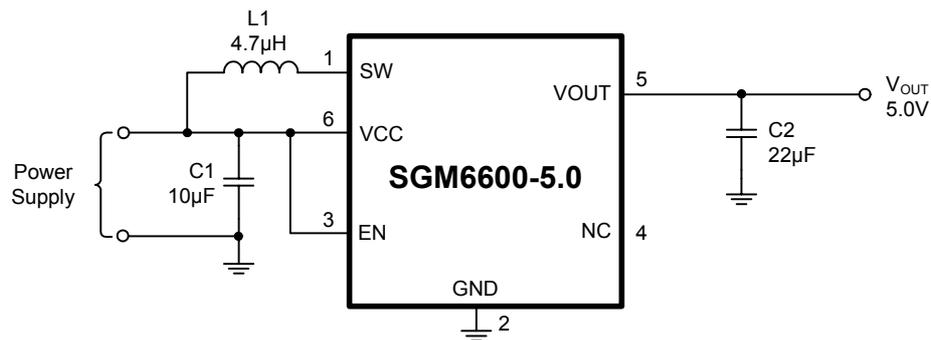


Figure 6b. Basic Application Circuit for the Fixed Output Versions

APPLICATION INFORMATION

Design Procedure

The SGM6600 DC/DC converter is intended for systems powered by a single-cell, up to triple-cell alkaline, NiCd, NiMH battery with a typical terminal voltage between 0.9V and 5.5V. They can also be used in systems powered by one-cell Li-ion or Li-polymer with a typical voltage between 2.5V and 4.2V.

Programming Output Voltage

In Figure 1, the output voltage of the SGM6600 DC/DC converter can be adjusted with an external resistor divider. The typical value of the voltage at the FB pin is 500mV. The maximum recommended value for the output voltage is 5.5V. R1 and R2 are calculated using Equation 1:

$$R1 = R2 \times \left(\frac{V_{OUT}}{V_{FB}} - 1 \right) = R2 \times \left(\frac{V_{OUT}}{500mV} - 1 \right) \quad (1)$$

R2 is recommended to be 100KΩ. For example, if an output voltage of 3.3V is needed, a 560KΩ resistor should be chosen for R1.

Inductor Selection

A boost converter normally requires two main passive components for storing energy during the conversion. A boost inductor and a storage capacitor at the output are required. To select the boost inductor, it is recommended to keep the possible peak inductor current below the current limit threshold of the power switch in the chosen configuration. The highest peak current through the inductor and the switch depends on the output load, the input (V_{CC}), and the output voltage (V_{OUT}). Estimation of the maximum average inductor current is done using Equation 2:

$$I_L = I_O \times \frac{V_{OUT}}{V_{CC} \times 0.8} \quad (2)$$

For example, for an output current of 75mA at 3.3V, at least 340mA of average current flows through the inductor at a minimum input voltage of 0.9V.

The second parameter for choosing the inductor is the desired current ripple in the inductor. Normally, it is advisable to work with a ripple of less than 20% of the average inductor current. A smaller ripple reduces the magnetic hysteresis losses in the inductor, as well as output voltage ripple and EMI. But in the same way, regulation time rises at load changes. In addition, a larger inductor increases the total system costs. With these parameters, it is possible to calculate the value for the inductor by using Equation 3:

$$L = \frac{V_{CC} \times (V_{OUT} - V_{CC})}{\Delta I_L \times f \times V_{OUT}} \quad (3)$$

Parameter f is the switching frequency and ΔI_L is the ripple current in the inductor, i.e., 40% ΔI_L . In this example, the desired inductor has the value of 4μH. With this calculated value and the calculated currents, it is possible to choose a suitable inductor. In typical applications, a 4.7μH inductance is recommended. The device has been optimized to operate with inductance values between 2.2μH and 10μH. Nevertheless, operation with higher inductance values may be possible in some applications. Detailed stability analysis is then recommended. Care must be taken because load transients and losses in the circuit can lead to higher currents as estimated in Equation 3. Also, the losses in the inductor caused by magnetic hysteresis losses and copper losses are a major parameter for total circuit efficiency.

Input Capacitor

At least a 10μF input capacitor is recommended to improve transient behavior of the regulator and EMI behavior of the total power supply circuit. A ceramic capacitor or a tantalum capacitor with a 100nF ceramic capacitor in parallel, placed close to the IC, is recommended.

APPLICATION INFORMATION

Output Capacitor

The major parameter necessary to define the output capacitor is the maximum allowed output voltage ripple of the converter. This ripple is determined by two parameters of the capacitor, the capacitance and the ESR. It is possible to calculate the minimum capacitance needed for the defined ripple, supposing that the ESR is zero, by using Equation 4:

$$C_{\text{MIN}} = \frac{I_{\text{O}} \times (V_{\text{OUT}} - V_{\text{CC}})}{f \times \Delta V \times V_{\text{OUT}}} \quad (4)$$

Parameter f is the switching frequency and ΔV is the maximum allowed ripple.

With a chosen ripple voltage of 10mV, a minimum capacitance of 4.5 μ F is needed. In this value range, ceramic capacitors are a good choice. The ESR and the additional ripple created are negligible. It is calculated using Equation 5:

$$\Delta V_{\text{ESR}} = I_{\text{O}} \times R_{\text{ESR}} \quad (5)$$

The total ripple is the sum of the ripple caused by the capacitance and the ripple caused by the ESR of the capacitor. Additional ripple is caused by load transients. This means that the output capacitor has to completely supply the load during the charging phase of the inductor.

The value of the output capacitance depends on the speed of the load transients and the load current during the load change. With the calculated minimum value of 4.5 μ F and load transient considerations, the recommended output capacitance value is in a 22 μ F range.

Care must be taken on capacitance loss caused by derating due to the applied dc voltage and the frequency characteristic of the capacitor. For example, larger form factor capacitors (in 1206 size) have their self resonant frequencies in the same frequency range as the SGM6600 operating frequency. So the effective capacitance of the capacitors used may be significantly lower. Therefore, the recommendation is to use smaller capacitors in parallel instead of one larger capacitor.

Layout Considerations

As for all switching power supplies, the layout is an important step in the design, especially at high-peak currents and high switching frequencies. If the layout is not carefully done, the regulator could show stability problems as well as EMI problems. Therefore, use wide and short traces for the main current path and for the power ground tracks. The input capacitor, output capacitor, and the inductor should be placed as close as possible to the IC. Use a common ground node for power ground and a different one for control ground to minimize the effects of ground noise. Connect these ground nodes at any place close to the ground pin of the IC.

The feedback divider should be placed as close as possible to the ground pin of the IC. To lay out the control ground, it is recommended to use short traces as well, separated from the power ground traces. This avoids ground shift problems, which can occur due to superimposition of power ground current and control ground current.

Thermal Information

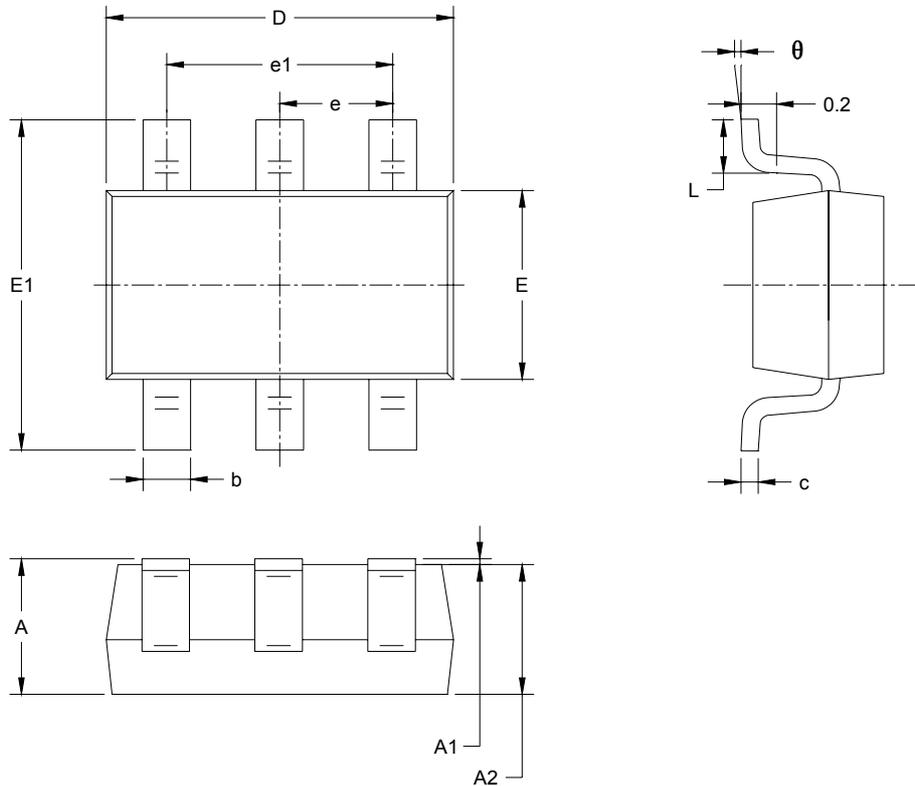
Implementation of integrated circuits in low-profile and fine-pitch surface-mount packages typically requires special attention to power dissipation. Many system-dependent issues such as thermal coupling, airflow, added heat sinks and convection surfaces, and the presence of other heat-generating components affect the power-dissipation limits of a given component.

Three basic approaches for enhancing thermal performance follow.

1. Improving the power dissipation capability of the PCB design.
2. Improving the thermal coupling of the component to the PCB.
3. Introducing airflow in the system

PACKAGE OUTLINE DIMENSIONS

SOT-23-6L



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.500	0.012	0.020
c	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.950 BSC		0.037 BSC	
e1	1.900 BSC		0.075 BSC	
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°

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